

**Amendments to the Claims:**

This listing of claims will replace all prior versions, and listings, of claims in the application:

**Listing of Claims:**

Claims 1-3 cancelled

Rule 1.126  
B'

4x

(Currently Amended) A COFDM demodulator, comprising:

a fast Fourier transform circuit for analyzing a received signal in a window corresponding to one symbol, each symbol carrying several phase and amplitude modulated carriers, some of which, shifted in frequency in a predetermined way from one symbol to the next one, form pilots;

a bidimensional filter for interpolating, from anchors corresponding to the pilots such as received from several consecutive symbols; the distortion undergone by each carrier;

means for correcting window shifting with respect to an optimal position; and

means for correcting each distortion according to window shifting corrections performed respectively for the symbol associated with the distortion and for the symbols associated with the anchors used to interpolate the distortion.

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(Previously Presented) The demodulator of claim 4 wherein the means for correcting the window shifting comprise a phase-locked loop synchronized on a correlation signal obtained by a correlation product between the received signal and this same signal delayed by one symbol, each symbol being preceded by a guard interval corresponding to a copy of the end of the symbol.

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(Previously Presented) The demodulator of claim 4 wherein each distortion is, in the frequency field after Fourier transform, a weighted sum of two anchors of the same position in a preceding symbol and in a following symbol, to which anchors have been added respective phases corresponding to the shiftings undergone by the analysis window for the preceding and following symbols, and to which anchors has been subtracted a phase

corresponding to the shifting undergone by the analysis window for the symbol associated with the distortion.

7/4. (Currently Amended) A COFDM demodulator with fast Fourier transform (FFT) analysis window displacement compensation comprising:

a reconstruction circuit configured to receive radio-transmitted signals in a window corresponding to one symbol, the symbol carrying a plurality phase and amplitude modulated carriers, one or more of the carriers are shifted in frequency in a predetermined way from one symbol to the next symbol and form pilots, the reconstruction circuit configured to extract the symbols and convert the symbols into digital signals;

an adjustment circuit and an associated phase-locked loop (PLL) circuit configured to receive the digital signals and determine and readjust the position of the corresponding windows;

an FFT circuit configured to perform a fast Fourier transform with the windows and output a transformed signal including complex coefficients;

a conversion circuit configured to receive a position signal from the PLL and to output a conversion signal that is corrected for distortion, the conversion circuit comprising first, second, and third analysis window shift value registers coupled to a first multiplexer; fourth, fifth, and sixth analysis window shift value registers coupled to a second multiplexer; the first and second multiplexers each having an output coupled to respective inputs of first and second adders; the first and second adders each having an output coupled to respective first and second multipliers; the first and second multipliers each having an output coupled to respective first and second polar-to-cartesian converters; and the first and second polar-to-cartesian converters each having an output coupled to respective second inputs of the first and second multipliers of the interpolation circuit;

a distortion interpolation circuit configured to receive the transformed signal and the conversion signal and to provide an interpolated distortion signal, the interpolation circuit comprising first, second, and third anchor input registers coupled to a first multiplexer; fourth, fifth, and sixth anchor input registers coupled to a second multiplexer; and first and second

multipliers each having inputs coupled respectively to the first and second multiplexers, and each further having an output coupled to a common adder; and

a correction circuit configured to receive the interpolated distortion signal and to output a corrected complex coefficients signal.

8. (Previously Presented) The demodulator of claim 7, further comprising a delay circuit coupled between the FFT circuit and the correction circuit and coupled in parallel with the distortion interpolation circuit.

9. (Previously Presented) The demodulator of claim 7 wherein the PLL is configured to be synchronized on a correlation signal obtained by a correlation product between the received radio-transmitted signals in a window and this same signal delayed by one symbol, each symbol being preceded by a guard interval corresponding to a copy of the end of the symbol.

10. (Previously Presented) The demodulator of claim 7 wherein the PLL comprises an accumulator that outputs the absolute position of the window with respect to a corresponding symbol.

11. (Previously Presented) The demodulator of claim 10 wherein the conversion circuit is configured to convert the absolute position received from the PLL into a form that is usable by the distortion interpolation circuit.

12. (Previously Presented) The demodulator of claim 7 wherein the PLL is configured to control the adjustment circuit.

13-14  
10-11. (Canceled)

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 12. (Currently Amended) The demodulator of claim 11~~7~~ wherein the first and second multipliers of the interpolation circuit comprise complex multipliers.

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 13. (Currently Amended) The demodulator of claim 11~~7~~ wherein the interpolation circuit is configured to calculate distortion according to the following:

$$d_{n,k} = \left(1 - \frac{s}{4}\right) A_{n-s,k} e^{j2\pi f_k (w_{n-s} - w_n)} + \frac{s}{4} A_{n+4-s,k} e^{j2\pi f_k (w_{n+4-s} - w_n)}$$

where

A is the received anchors,

S is equal to  $(n \bmod 4 - k/3 \bmod 4) \bmod 4$ ,

n is the symbol number,

k is the window position,

$f_k$  is the frequency corresponding to position k, and

w is the absolute window position expressed in time units of the associated window.

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 14. (Currently Amended) A method of fast Fourier transform (FFT) analysis window displacement compensation in a COFDM modulator, comprising:

receiving a radio-transmitted signal in a window corresponding to one symbol, the symbol carrying a plurality of phase and amplitude modulated carriers, one or more of the carriers are shifted in frequency from one symbol to the next symbol to form pilots, and extracting these symbols and converting these symbols into digital signals;

receiving the digital signals and determining and readjusting the position of the windows;

receiving the windows and performing a fast Fourier transform with the windows and outputting a transformed signal that includes complex coefficients;

receiving a position signal and outputting a conversion signal that is corrected for distortion;

receiving the transformed signal and the conversion signal, including anchors corresponding to pilots received from several consecutive symbols, and providing an interpolated signal of distortion undergone by each carrier; and

receiving the interpolated signal and outputting a corrected coefficient signal.

<sup>18</sup><sub>18</sub>. (Previously Presented) The method of claim <sup>17</sup><sub>14</sub>, further comprising receiving the transformed signal and outputting a delayed transformed signal.

<sup>19</sup><sub>18</sub>. (Previously Presented) The method of claim <sup>17</sup><sub>14</sub>, further comprising generating a phase-locked loop signal synchronized on a correlation signal obtained by a correlation product between the received signal and this same signal delayed by one symbol, each symbol being preceded by a guard interval corresponding to a copy of the end of the symbol.

<sup>20</sup><sub>17</sub>. (Previously Presented) The method of claim <sup>17</sup><sub>14</sub>, comprising calculating distortion according to the following:

$$d_{n,k} = \left(1 - \frac{s}{4}\right) A_{n-s,k} e^{j2\pi f_k (w_{n-s} - w_n)} + \frac{s}{4} A_{n+4-s,k} e^{j2\pi f_k (w_{n+4-s} - w_n)}$$

where

A is the received anchors,

S is equal to (n modulo 4 - k/3 modulo 4) modulo 4,

n is the symbol number,

k is the window position,

$f_k$  is the frequency corresponding to a position k, and

w is the absolute window position expressed in time units of the associated window.